

# Exact Linearization Control for Twin Rotor MIMO System

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**Abstract:**

*Twin Rotor MIMO System (TRMS) is a non-linear object of which the dynamic behavior resembles that of a helicopter. The TRMS has been significantly used to verify control algorithms' performance and it attracts attention of scholars in the area of modeling and control. In this paper, an exact linearization controller is designed for the TRMS according to the Euler Lagrange model. The outputs of the controller are compared to the reference inputs. All related formulas are analyzed and experimental results are shows that TRMS tracks the desired trajectory accurately.*

**Keywords:** Exact linearization, Twin Rotor MIMO System

## I. INTRODUCTION

Twin Rotor multiple input - multiple output (MIMO) System (TRMS), as shown in Figure 1-3, is an experimental system developed by Feedback Instrument Ltd (Feedback Co., 1998) for MIMO experiments. TRMS consists of two rotors placed on a beam together with a counterbalance. The whole unit is attached to a tower allowing for safe helicopter control experiments. The movements in the vertical plane and in the horizontal plane are implemented by a vertical rotor (main rotor) and a horizontal rotor (tail rotor) respectively. The above rotors are driven by DC motors.

TRMS's dynamic characteristics are similar to that of a helicopter and they are nonlinear systems and have a significant cross coupling between two rotors. It is challenging for scientists to design a controller with acceptable outcome. Domestically and internationally, there have been a large number of publications regarding controllers for TRMS. Nevertheless, according to their experimental results, the outputs have mostly tracked desired trajectories but significant errors have been recognized. Therefore, in this paper, a design method based on the Euler-Lagrange model of TRMS will be proposed.



Figure 1. Twin Rotor MIMO System

The rest of the paper is organized as follows. The object's model and the controller's design are addressed in Section II. Simulation and evaluation will be demonstrated in Section III. Finally, conclusions and also future work will be mentioned in Section IV. The list of key notations used in this paper are provided in Table I

Table I Key Notations

Symbol	Unit	Feature Exaction Full name
$\alpha_{vh}$	rad	Angular position of the beam in vertical/horizontal plane
$\alpha_{mt}$	rad	Angular position of the main/tail propeller
$g$	m/s <sup>2</sup>	Gravity acceleration
$m$	kg	Mass
$J_1$	kgm <sup>2</sup>	Moment of Inertia of the beam
$m_{T1}$	kg	Total mass of the beam
$l_{T1}$	m	Centre of gravity of the beam
$m_t$	kg	Mass of the tail part of the beam
$m_{tr}$	kg	Mass of the horizontal rotor (tail rotor)
$m_{ts}$	kg	Mass of the tail shield
$m_m$	kg	Mass of the main part of the beam
$m_{mr}$	kg	Mass of the vertical rotor (the main rotor)
$m_{ms}$	kg	Mass of the main shield
$l_t$	m	Length of the tail part of the beam
$l_m$	m	Length of the main part of the beam
$r_{mts}$	m	Radius of the main/tail shield
$r_{mm/t}$	m	Radius of the main/tail rotor
$J_2$	kgm <sup>2</sup>	Moment of Inertia of the counterbalance beam

$m_b$	kg	Mass of the counterbalance beam
$m_{T2}$	kg	Total mass of the counterbalance beam
$m_{cb}$	kg	Mass of the counter-weight
$l_{T2}$	m	Centre of gravity of the counterbalance beam
$l_b$	m	Length of the counterbalance beam
$l_{cb}$	m	Distance from the counter-weight to the pivot
$r_{cb}$	m	Radius of the counterbalance
$L_{cb}$	m	Length of the counterbalance
$J_3$	kgm <sup>2</sup>	Moment of Inertia of the pivot
$J_4$	kgm <sup>2</sup>	Moment of Inertia of the rear part of the pivot
$m_h$	kg	Mass of the pivot
$m_{hl}$	kg	Mass of the rear part of the pivot
$h$	m	Length of the pivot
$h_l$	m	chiều dài phần sau của chốt quay
$J_{mm}$	kgm <sup>2</sup>	Moment of Inertia of motors
$J_{m/p}$	kgm <sup>2</sup>	Moment of Inertia of main/tail propellers
$\omega_{m/t}$	rad/s	Rotational speed of main/tail rotor
$J_{m/tr}$	kgm <sup>2</sup>	Moment of Inertia of mail/tail rotor
$H$	m	High from the base to the pivot
$k_g$		Gyroscopic momentum parameter
$M_v$	Nm	Total moment in vertical plane
$M_h$	Nm	Total moment in horizontal plane
$M_{m/t}$	Nm	Total moment exerted on main/tail rotor
$B_{m/tr}$	kgm <sup>2</sup> /s	Viscous friction constant of main/tail motors.
$B_{v/h}$	kgm <sup>2</sup> /s	Viscous friction constant of the pivot in vertical/horizontal plane.
$F_{v/h}$	Nm	Sliding friction of the pivot in vertical/horizontal plane.
$\tau_{m/t}$	Nm	Electromagnetic moment of motor for vertical rotor/ horizontal motor

II. CONTROL DESIGN

Euler-Lagrange Equation

Lagrange function of TRMS is concluded from the sum of potential and kinetic energies [2] as follows.

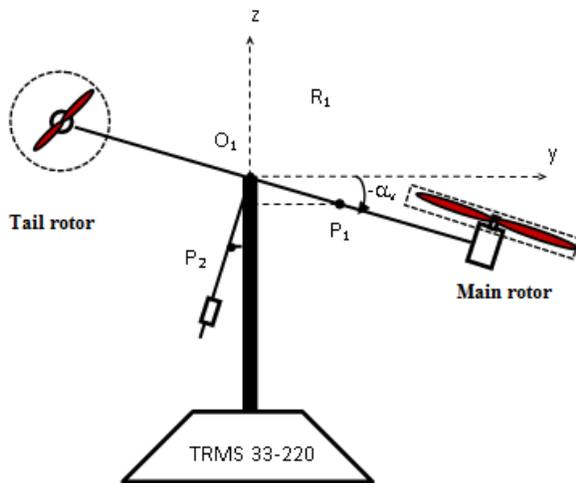


Figure 2. Vertical Position  $\alpha_v$  of TRMS

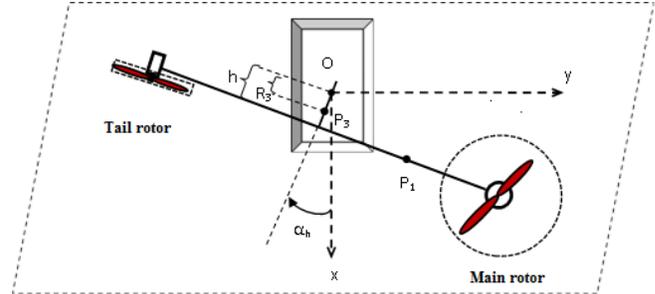


Figure 3. Horizontal Position  $\alpha_h$  of TRMS

$$L = \frac{a_1}{2} \dot{\alpha}_v^2 + \frac{1}{2} (a_5 + a_4 \cos^2 \alpha_v) \dot{\alpha}_h^2 + a_6 \omega_t \dot{\alpha}_v + a_7 \omega_m \dot{\alpha}_h \cos \alpha_v + \frac{a_6}{2} \omega_t^2 + \frac{a_7}{2} \omega_m^2 - b_1 \sin \alpha_v + (a_2 \sin \alpha_v - a_3 \cos \alpha_v) \dot{\alpha}_v \dot{\alpha}_h + b_2 \cos \alpha_v$$

where

$$a_1 = J_1 + J_2 + J_{tr}; a_2 = m_{T2} l_{T2} h; a_3 = m_{T1} l_{T1} h; a_4 = J_1 + J_{mr} - J_2; b_1 = m_{T1} l_{T1} g; b_2 = m_{T2} l_{T2} g; a_5 = J_3 + J_4 + m_{T2} h^2 + m_{T1} h^2 + J_2; a_6 = J_{tr}; a_7 = J_{mr}; \dot{\alpha}_t = \omega_t; \dot{\alpha}_m = \omega_m$$

$$J_1 = \left( \frac{1}{3} m_t + m_{tr} + m_{ts} \right) l_t^2 + \left( \frac{1}{3} m_m + m_{mr} + m_{ms} \right) l_m^2 + \frac{1}{2} m_{ms} r_{ms}^2 + m_{ts} r_{ts}^2 + \frac{1}{2} m_{tr} r_{tr}^2 + \frac{1}{2} m_{mr} r_{mr}^2; m_{T1} = m_t + m_{tr} + m_{ts} + m_m + m_{mr} + m_{ms}; l_{T1} = \frac{(m_m / 2 + m_{m_t} + m_{m_s}) l_m - (m_t / 2 + m_{tr} + m_{ts}) l_t}{m_{T1}}; J_2 = \frac{1}{3} m_b l_b^2 + m_{cb} l_{cb}^2 + \frac{1}{4} m_{cb} r_{cb}^2 + \frac{1}{12} m_{cb} L_{cb}^2; m_{T2} = m_b + m_{cb}; l_{T2} = \frac{m_b l_b / 2 + m_{cb} l_{cb}}{m_{T2}}; J_4 = \frac{m_{h1} h_1^2}{3}; J_4 = \frac{m_{h1} h_1^2}{3}$$

Apply Lagrange's equation:

$$\frac{d}{dt} \left( \frac{\partial L}{\partial \dot{q}} \right) - \frac{\partial L}{\partial q} = M_{ex},$$

$$\text{where } q = [\alpha_v \quad \alpha_h \quad \alpha_m \quad \alpha_t]^T.$$

Then, the dynamic Euler-Lagrange of TRMS can be written as follows.

$$M \ddot{q} + C \dot{q} + G = M_{ex},$$

where

$$\ddot{q} = [\ddot{\alpha}_v \quad \ddot{\alpha}_h \quad \ddot{\alpha}_m \quad \ddot{\alpha}_t]^T, \dot{q} = [\dot{\alpha}_v \quad \dot{\alpha}_h \quad \dot{\alpha}_m \quad \dot{\alpha}_t]^T$$

$$G = [b_2 \sin \alpha_v + b_1 \cos \alpha_v \quad 0 \quad 0 \quad 0]^T$$

$$M_{ex} = [M_v \quad M_h \quad M_m \quad M_t]^T$$

and

$$M = \begin{bmatrix} a_1 & z_1 & 0 & a_6 \\ z_2 & z_3 & a_7 \cos \alpha_v & 0 \\ 0 & a_7 \cos \alpha_v & a_7 & 0 \\ a_6 & 0 & 0 & a_6 \end{bmatrix}, \quad (4)$$

where

$$z_1 = a_2 \sin \alpha_v - a_3 \cos \alpha_v, z_2 = a_2 \sin \alpha_v - a_3 \cos \alpha_v$$

$$z_3 = a_5 + a_4 \cos^2 \alpha_v$$

$$C = \begin{bmatrix} 0 & n_1 & 0 & 0 \\ n_2 & -a_4 \sin \alpha_v \cos \alpha_v \dot{\alpha}_v & 0 & 0 \\ 0 & -a_7 \dot{\alpha}_v \sin \alpha_v & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}, \quad (5)$$

where

$$n_1 = \sin \alpha_v (a_4 \dot{\alpha}_h \cos \alpha_v + a_7 \omega_m)$$

$$n_2 = -a_4 \sin \alpha_v \cos \alpha_v \dot{\alpha}_h - a_7 \omega_m \sin \alpha_v + (a_2 \cos \alpha_v + a_3 \sin \alpha_v) \dot{\alpha}_v$$

$$M_v = l_m k_{fv} \dot{\alpha}_m \left| \dot{\alpha}_m \right| w - k_{rv} \dot{\alpha}_i \left| \dot{\alpha}_i \right| - B_v \dot{\alpha}_v - F_v \text{sign}(\dot{\alpha}_v) - k_{gm} k_{fv} \dot{\alpha}_m \left| \dot{\alpha}_m \right| w \dot{\alpha}_h \cos \alpha_v$$

(6)

where

$$w = \frac{1}{1 - \left( \frac{r_{mr}}{4(H + l_m \sin \alpha_v)} \right)^2}$$

$$M_h = l_t k_{fh} \dot{\alpha}_i \left| \dot{\alpha}_i \right| \cos \alpha_v - k_{tm} \dot{\alpha}_m \left| \dot{\alpha}_m \right| \cos \alpha_v - B_h \dot{\alpha}_h - F_h \text{sign}(\dot{\alpha}_h) - C_c (\alpha_h - \alpha_{h0})$$

(7)

$$M_m = \tau_m - \text{sign}(\dot{\alpha}_m) k_{rv} \dot{\alpha}_m^2 - B_{mr} \dot{\alpha}_m \quad (8)$$

$$M_t = \tau_t - \text{sign}(\dot{\alpha}_t) k_{th} \dot{\alpha}_t^2 - B_{tr} \dot{\alpha}_t. \quad (9)$$

Consequently, the above Euler-Lagrange model shows that TRMS lacks actuators with 2 inputs and 4 outputs.

### A. Control Design

Firstly, equation (10) can be rewritten as follows:

$$M \ddot{q} + C \dot{q} + G = M_{ex}$$

or

$$\begin{cases} M_{11} \ddot{q}_1 + M_{12} \ddot{q}_2 + C_{11} \dot{q}_1 + C_{12} \dot{q}_2 + \phi_1 = 0 \\ M_{21} \ddot{q}_1 + M_{22} \ddot{q}_2 + C_{21} \dot{q}_1 + C_{22} \dot{q}_2 + \phi_2 = \begin{bmatrix} \tau_m \\ \tau_t \end{bmatrix} \end{cases} \quad (12)$$

where

$$q_1 = [\alpha_v \quad \alpha_h]^T, q_2 = [\alpha_m \quad \alpha_t]^T$$

$$M_{11} = \begin{bmatrix} a_1 & a_2 \sin \alpha_v - a_3 \cos \alpha_v \\ a_2 \sin \alpha_v - a_3 \cos \alpha_v & a_5 + a_4 \cos^2 \alpha_v \end{bmatrix}$$

$$M_{12} = \begin{bmatrix} 0 & a_6 \\ a_7 \cos \alpha_v & 0 \end{bmatrix}, M_{21} = \begin{bmatrix} 0 & a_7 \cos \alpha_v \\ a_6 & 0 \end{bmatrix}$$

$$M_{22} = \begin{bmatrix} a_7 & 0 \\ 0 & a_6 \end{bmatrix}$$

and

$$C_{11} = \begin{bmatrix} 0 & \sin \alpha_v (a_4 \dot{\alpha}_h \cos \alpha_v + a_7 \omega_m) \\ x & -a_4 \sin \alpha_v \cos \alpha_v \dot{\alpha}_v \end{bmatrix}$$

$$x = -a_4 \sin \alpha_v \cos \alpha_v \dot{\alpha}_h - a_7 \omega_m \sin \alpha_v + (a_2 \cos \alpha_v + a_3 \sin \alpha_v) \dot{\alpha}_v$$

$$C_{12} = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix}, C_{21} = \begin{bmatrix} 0 & -a_7 \dot{\alpha}_v \sin \alpha_v \\ 0 & 0 \end{bmatrix}, C_{22} = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix}$$

and

$$\phi_1 = \begin{bmatrix} b_2 \sin \alpha_v + b_1 \cos \alpha_v - M_v \\ -M_h \end{bmatrix}$$

$$\phi_2 = \begin{bmatrix} \text{sign}(\omega_m) k_{rv} \omega_m^2 + B_{mr} \omega_m \\ \text{sign}(\omega_t) k_{th} \omega_t^2 + B_{tr} \omega_t \end{bmatrix}$$

From (4), we have

$$\begin{cases} M_{11} \ddot{q}_1 + M_{12} \ddot{q}_2 + C_{11} \dot{q}_1 + C_{12} \dot{q}_2 + \phi_1 = 0 \\ M_{21} \ddot{q}_1 + M_{22} \ddot{q}_2 + C_{21} \dot{q}_1 + C_{22} \dot{q}_2 + \phi_2 = \begin{bmatrix} \tau_m \\ \tau_t \end{bmatrix} \end{cases} \quad (13)$$

or

$$\begin{cases} \ddot{q}_2 = -M_{12}^{-1} (M_{11} \ddot{q}_1 + C_{11} \dot{q}_1 + C_{12} \dot{q}_2 + \phi_1) \\ M_{21} \ddot{q}_1 + M_{22} \ddot{q}_2 + C_{21} \dot{q}_1 + C_{22} \dot{q}_2 + \phi_2 = \tau \end{cases} \quad (14)$$

$$\text{where } \tau = \begin{bmatrix} \tau_m \\ \tau_t \end{bmatrix}.$$

A control problem for TRMS is that the first joint variable  $q_1$  has to approach the reference trajectory  $q_{1r}$  regardless of the second joint parameter  $q_2$ . It can be seen that the first controller, namely inner loop, can be expressed in the following the equations as follows.

$$\begin{aligned} \tau &= M_{21} \ddot{q}_1 + C_{21} \dot{q}_1 + \phi_2 \\ &\quad - M_{22} M_{12}^{-1} (M_{11} \ddot{q}_1 + C_{11} \dot{q}_1 + C_{12} \dot{q}_2 + \phi_1) \end{aligned} \quad (15)$$

$$\begin{aligned} & (M_{21} - M_{22} M_{12}^{-1} M_{11}) \ddot{q}_1 + C_{21} \dot{q}_1 + C_{22} \dot{q}_2 \\ & \quad + \phi_2 - M_{22} M_{12}^{-1} (C_{11} \dot{q}_1 + C_{12} \dot{q}_2 + \phi_1) \end{aligned} \quad (16)$$

where

$$\begin{cases} \varphi(q, \dot{q}) = C_{21} \dot{q}_1 + C_{22} \dot{q}_2 + \phi_2 - M_{22} M_{12}^{-1} (C_{11} \dot{q}_1 + C_{12} \dot{q}_2 + \phi_1) \\ D = M_{21} - M_{22} M_{12}^{-1} M_{11} \end{cases}$$

The closed loop system will have a joint variable set that is derived from the controller (17) and the model (18) as follows:

$$D \ddot{q}_1 = D v \Leftrightarrow \ddot{q}_1 = v . \tag{19}$$

In other words, the inner controller in Equation (20) has linearized  $q_1$  in Equation (21) in the whole joint variable domain. Therefore, the inner controller can be rewritten as follows.

$$\tau = D v + \varphi(q, \dot{q}) . \tag{22}$$

The controller in Equation (9) is called the exact linearization controller. Obviously, despite the  $q_1$  linearization, the system is unstable because of the second order integral in Equation (23).

After that, in order to apply the unstable  $q_1$  to the desired values  $q_{1r}$ , the outer loop controller will be used.

$$v = \ddot{q}_{1r} + K_1 e_1 + K_2 \dot{e}_1 , \tag{24}$$

where  $e_1 = q_{1r} - q_1$  . \tag{25}

$e_1$  denotes the tracking error;  $K_1, K_2$ , are two arbitrary symmetric positive definite matrices. The outer loop controller in (26) probably makes the tracking error  $e_1$  decrease to zero. From (27) and (28), we have:

$$\ddot{q}_1 = \ddot{q}_{1r} + K_1 e_1 + K_2 \dot{e}_1 , \tag{29}$$

and

$$\underline{0} = \ddot{e}_1 + K_1 e_1 + K_2 \dot{e}_1 , \tag{30}$$

and

$$\begin{pmatrix} \dot{e}_1 \\ \ddot{e}_1 \end{pmatrix} = \begin{pmatrix} 0 & I \\ -K_1 & -K_2 \end{pmatrix} \begin{pmatrix} e_1 \\ \dot{e}_1 \end{pmatrix} = K \begin{pmatrix} e_1 \\ \dot{e}_1 \end{pmatrix} , \tag{31}$$

where

$$K = \begin{pmatrix} 0 & I \\ -K_1 & -K_2 \end{pmatrix} . \tag{32}$$

is the Hurwitz matrix.

Finally, the addition of the outer loop controller leads the system to having the total controller as shown in Equation (33).

$$\tau = D (\ddot{q}_{1r} + K_1 e_1 + K_2 \dot{e}_1) + \varphi(q, \dot{q}) . \tag{34}$$

Figure 4 illustrates the diagram of the tracking system according to the exact linearization of  $q_1$ .

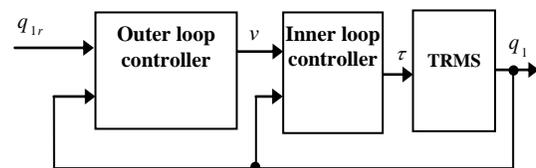


Figure 4. TRMS Control Diagram

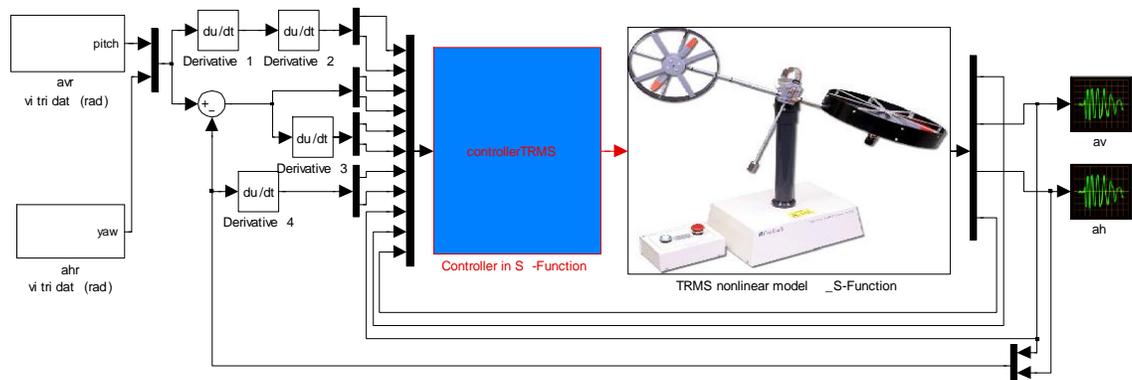


Figure 5. TRMS Control System Simulation

### III. SIMULATION RESULTS

With the aim of verifying the performance of the proposed controller, MATLAB-SIMULINK has been used to conduct simulations according to parameters in Table II. The model of TRMS and controller are implemented via S-function. For the sake of simplicity, there will be one S-function for both the outer loop controller and the inner loop controller. The output position  $\alpha_{v/h} = av/h$  ( $\alpha_v$ =pitch angle,  $\alpha_h$ =yaw angle) and the reference angle  $q_{1r} = [avr \ ahr]$ . In order to observe effects of the cross reaction, the simulation with the 2 degree of freedom model and 2 reference inputs have been created. Let's choose  $K1=1.I$  and

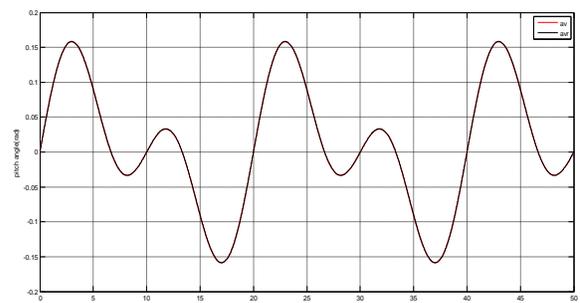


Figure 6. Vertical Position  $\alpha_v$  with the Reference Input  $0.09 \sin(0.6283t) + 0.09 \sin(0.3142t)$

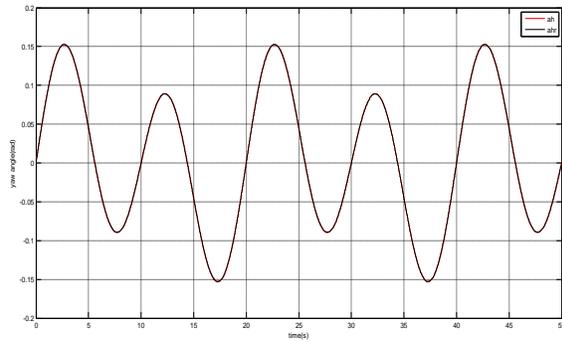


Figure 7. Horizontal Position  $\alpha_h$  with the Reference Input:  $0.12 \sin(0.6283t) + 0.045 \sin(0.3142t)$

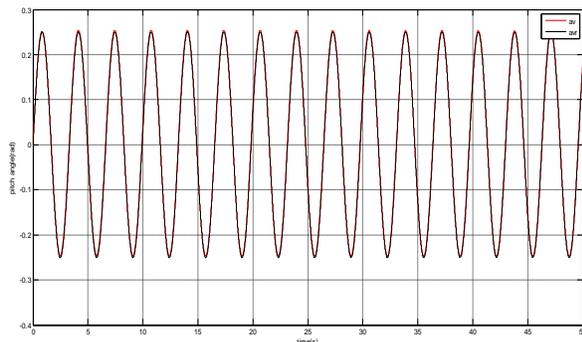


Figure 8. Vertical Position  $\alpha_v$  with the Reference Input:  $0.25 \sin(1.9t)$

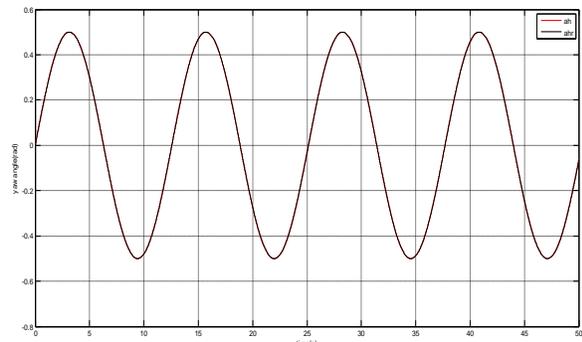


Figure 9. Horizontal Position  $\alpha_h$  with the Reference Input:  $0.5 \sin(0.2t)$

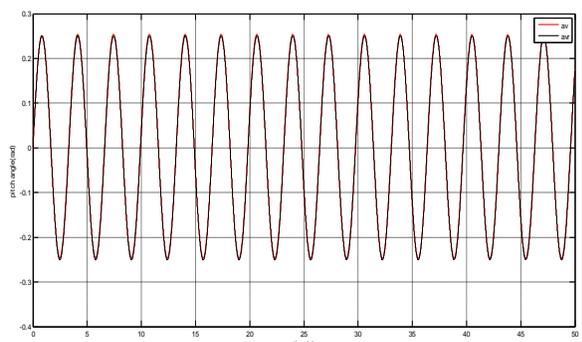


Figure 10. Vertical Position  $\alpha_v$  with the Reference Input:  $0.25 \sin(1.9t)$

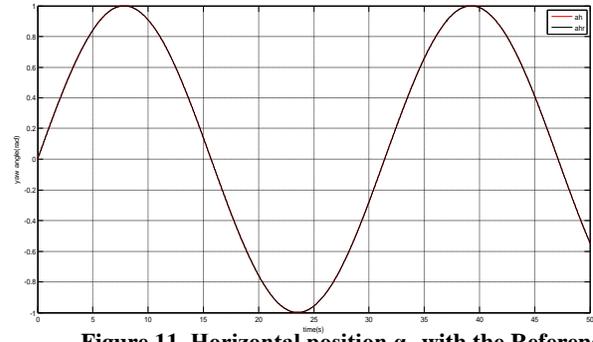


Figure 11. Horizontal position  $\alpha_h$  with the Reference input:  $1 \sin(0.5t)$

Table II TRMS Parameters

$l_t$	0.282 m	$k_g$	0.05
$l_m$	0.254 m	$h$	$6e-2$ m
$l_b$	0.265 m	$h_l$	0.02 m
$l_{cb}$	0.25 m	$m_{h1}$	0.05 kg
$r_{ms}$	0.155 m	$m_h$	0.09 kg
$r_{ts}$	0.1 m	$g$	$9.81 \text{ m/s}^2$
$m_{tr}$	0.221 kg	$L_c$	$3e-2$ m
$m_{mr}$	0.236 kg	$r_{cb}$	$1e-2$ m
$m_{cb}$	0.068 kg	$H$	0.5 m
$m_m$	0.014 kg	$r_{mt}$	0.007 m
$m_t$	0.015 kg	$r_{mm}$	0.007 m
$m_b$	0.022 kg	$m_{mrr}$	0.042 kg
$m_{ts}$	0.119 kg	$m_{trr}$	0.016 kg
$m_{ms}$	0.219 kg	$k_{chp}$	0.00854
$J_{mr}$	$21.624e-5 \text{ kgm}^2$	$J_{tr}$	$3.1432e-5 \text{ kgm}^2$
$B_{mr}$	$4.5e-5 \text{ kgm}^2/\text{s}$	$B_{tr}$	$2.3e-5 \text{ kgm}^2/\text{s}$
$k_{tv}$	$23.03e-6$	$k_{th}$	$10e-6$
$B_v$	$0.6e-2 \text{ Nms/rad}$	$F_v$	$0.1e-2 \text{ Nms/rad}$
$B_h$	$0.1 \text{ Nms/rad}$	$F_h$	$0.01 \text{ Nms/rad}$
$C_c$	$0.016 \text{ Nm/rad}$	$\alpha_{h0}$	$-0.4602 \text{ rad}$

#### IV. CONCLUSIONS AND FUTURE WORK

In this paper, the 2-degree of freedom movement of TRMS has been discussed. The mathematical model of TRMS based on Euler Lagrange equations has been built by using MATLAB/SIMULINK. The controller is designed to be able to work in vertical and horizontal planes. Outputs of the controller have been compared to the

reference inputs. It is clear that TRMS tracks the desired trajectory accurately. In our future work, the application of the controller for a real object would be implemented.

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